

CEU

*Universidad  
San Pablo*

## UNIT 4: Optimal Filtering

Gabriel Caffarena Fernández  
3<sup>rd</sup> Year Biomedical Engineering Degree  
EPS – Univ. San Pablo – CEU

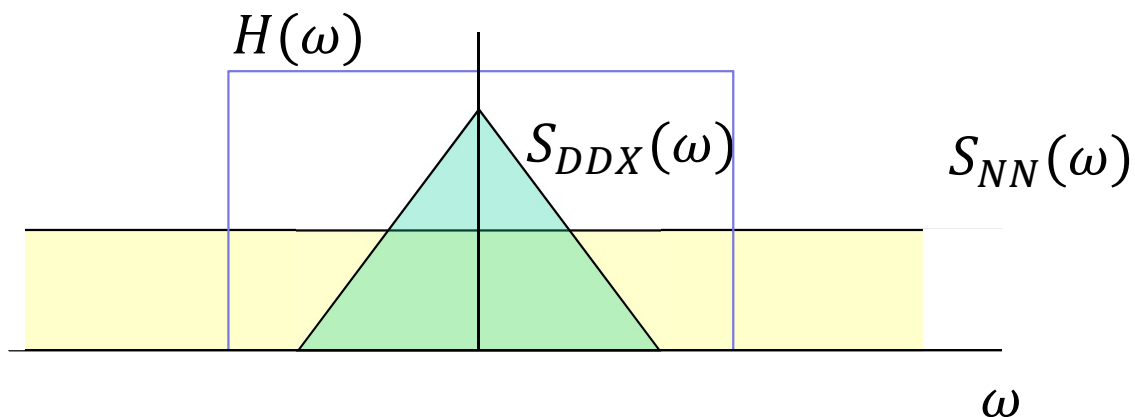
(based on “Biomedical Signal Analysis”, 2nd edition, © Willey 2015)

# Noise filtering

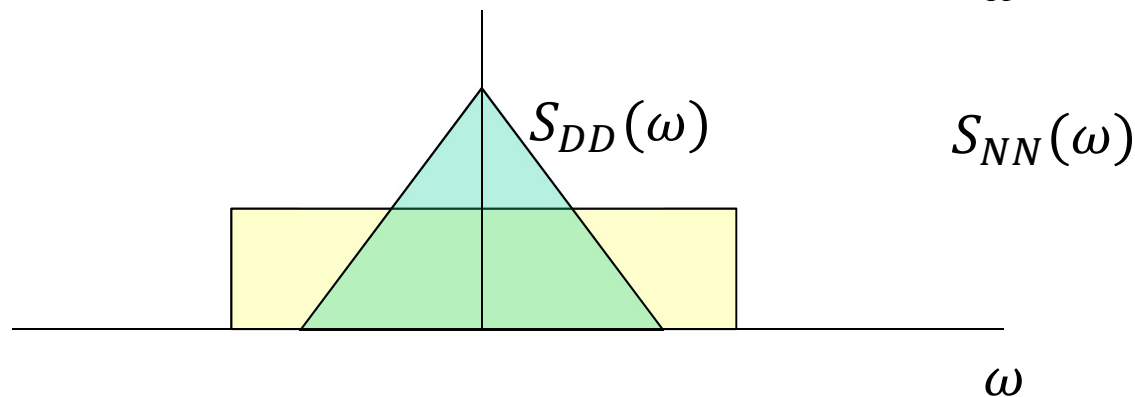
- A common technique to reduce the effect of noise is filtering with a band pass equal to the frequency band of the signal of interest. The effect is a reduction in the signal to noise ratio (SNR)

$$SNR = 10 \log \left( \frac{P_D}{P_N} \right)$$

SNR is low



SNR is increased



# Noise filtering

- In order to do so, we have to select the following:
  - Type of filter: **FIR** or **IIR** (digital domain)
  - Order of filter
  - Location of poles and zeros
- The previous decisions lead to many different filter properties that can be desired or not:
  - Flat or rippled band pass
  - Phase distortion or linear phase
  - Smooth or abrupt transition band
- However, it is possible to design a digital filter considering the **statistical properties** of signal and noise and **maximizing SNR**

# Optimal filtering

- **Problem:**

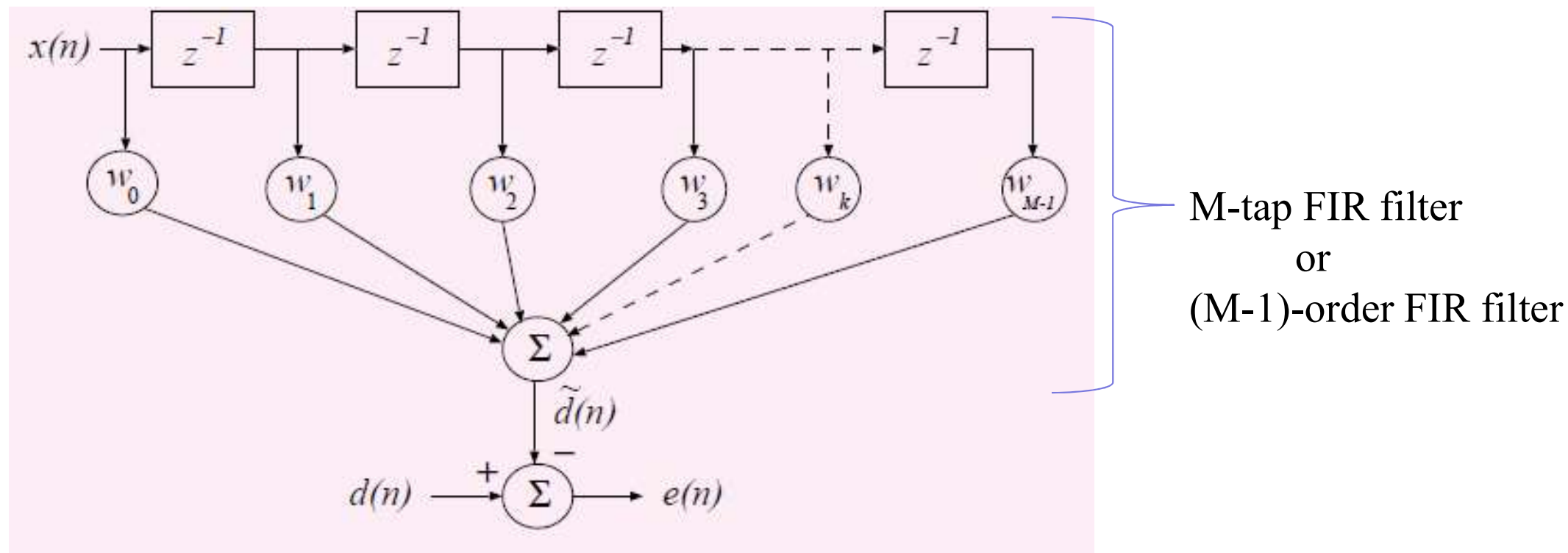
- Design an optimal filter to remove noise from a signal, given that the signal and noise processes are random, independent and stationary.
- We assume the “desired” or ideal characteristics of the uncorrupted signal as well as characteristics of the noise to be known.

- **Solution:**

- Wiener filter theory provides a means to optimize filter parameters with reference to a performance criterion.
- The output is guaranteed to be the best achievable result

# The Wiener filter

- $\mathbf{d}(n)$  is the desired signal
- $\mathbf{x}(n)$  is  $\mathbf{d}(n)$  plus noise
- $\mathbf{x}(n)$  is filtered producing an estimate  $\hat{\mathbf{d}}(n)$  of the desired signal
- Our goal is to minimize the error  $\mathbf{e}(n) = \mathbf{d}(n) - \hat{\mathbf{d}}(n)$



# The Wiener filter

- The filter has  $M$  tap weights  $w_i, i \in \{0, 1, 2, \dots, M - 1\}$ , so for an FIR filter the impulse

$$\text{response is } h(n) = \begin{cases} w_n, n \in [0, M - 1] \\ 0, \text{otherwise} \end{cases}$$

- The output of the filter is the convolution of the impulse response with  $\mathbf{x}(n)$

$$\hat{d}(n) = h(n) * x(n) = \sum_{k=0}^{M-1} h(n)x(n - k) = \sum_{k=0}^{M-1} w_k x(n - k)$$

- Let's introduce vectors to use more compact expressions

$$\vec{w} = [w_0, w_1, w_2, \dots, w_{M-1}]^T \quad \text{coefficient vector}$$

$$\vec{x}(n) = [x(n), x(n - 1), \dots, x(n - (M - 1))]^T \quad \text{delayed input vector}$$

**NOTE** that they are column vectors

- So now we can rewrite  $\hat{d}(n)$  and  $e(n)$

$$\hat{d}(n) = \sum_{k=0}^{M-1} w_k x(n - k) = \vec{w}^T \vec{x}(n)$$

$$e(n) = d(n) - \hat{d}(n) = d(n) - \vec{w}^T \vec{x}(n)$$

# The Wiener filter

- The criterion used to optimize the error is to minimize the **Mean Squared Error**

$$\text{MSE} = E[e^2(n)]$$

$$= E \left[ \left( d(n) - \hat{d}(n) \right)^2 \right]$$

$$= E \left[ \left( d(n) - \vec{w}^T \vec{x}(n) \right) \left( d(n) - \vec{x}(n)^T \vec{w} \right) \right]$$

$$= E[d^2(n)] - \vec{w}^T E[\vec{x}(n)d(n)] - E[d(n)\vec{x}(n)^T] \vec{w} + \vec{w}^T E[\vec{x}(n)\vec{x}(n)^T] \vec{w}$$

# The Wiener filter

- Let's now add vector  $\overrightarrow{R_{XD}}$

$$\text{MSE} = E[e^2(n)]$$

$$= E[d^2(n)] - \overrightarrow{w}^T E[\vec{x}(n)d(n)] - E[d(n)\vec{x}(n)^T]\overrightarrow{w} + \overrightarrow{w}^T E[\vec{x}(n)\vec{x}(n)^T]\overrightarrow{w}$$

$$\begin{aligned} E[\vec{x}(n)d(n)] &= [E[x(n)d(n)], E[x(n-1)d(n)], \dots, E[x(n-(M-1))d(n)]]^T \\ &= [R_{XD}(0), R_{XD}(1), \dots, R_{XD}(M-1)]^T = \overrightarrow{R_{XD}} \end{aligned}$$

$$\begin{aligned} E[d(n)\vec{x}(n)^T] &= [E[d(n)x(n)], E[d(n)x(n-1)], \dots, E[d(n)x(n-(M-1))d(n)]] \\ &= [R_{DX}(0), R_{DX}(-1), \dots, R_{DX}(-(M-1))] = \\ &= [R_{XD}(0), R_{XD}(1), \dots, R_{XD}(M-1)] = \overrightarrow{R_{XD}}^T \end{aligned}$$



# The Wiener filter

- Let's now add vector  $\overrightarrow{R_{XD}}$

$$\text{MSE} = E[e^2(n)]$$

$$= E[d^2(n)] - \overrightarrow{w}^T E[\vec{x}(n)d(n)] - E[d(n)\vec{x}(n)^T]\overrightarrow{w} + \overrightarrow{w}^T E[\vec{x}(n)\vec{x}(n)^T]\overrightarrow{w}$$

$$= E[d^2(n)] - \overrightarrow{w}^T \overrightarrow{R_{XD}} - \overrightarrow{R_{XD}}^T \overrightarrow{w} + \overrightarrow{w}^T E[\vec{x}(n)\vec{x}(n)^T]\overrightarrow{w}$$

$$E[\vec{x}(n)d(n)] = [R_{XD}(0), R_{XD}(1), \dots, R_{XD}(M-1)]^T = \overrightarrow{R_{XD}}$$

$$E[d(n)\vec{x}(n)^T] = [R_{DX}(0), R_{DX}(-1), \dots, R_{DX}(-(M-1))] = \overrightarrow{R_{XD}}^T$$

- It is trivial to check that  $\overrightarrow{w}^T \overrightarrow{R_{XD}} = \overrightarrow{R_{XD}}^T \overrightarrow{w}$ , so

$$\text{MSE} = E[e^2(n)] = E[d^2(n)] - 2\overrightarrow{w}^T \overrightarrow{R_{XD}} + \overrightarrow{w}^T E[\vec{x}(n)\vec{x}(n)^T]\overrightarrow{w}$$

$R_{XX}?$   
↓

# The Wiener filter

- And finally, matrix  $\mathbf{R}_{XX}$

$$\begin{aligned} \text{MSE} &= E[e^2(n)] \\ &= E[d^2(n)] - 2\vec{w}^T \overrightarrow{R_{XD}} + \vec{w}^T \mathbf{R}_{XX} \vec{w} \end{aligned}$$

$$\mathbf{R}_{XX} = E[\vec{x}(n)\vec{x}(n)^T]$$

$$= \begin{pmatrix} E[x(n)x(n)] & E[x(n)x(n-1)] & \dots & E[x(n)x(n-M+1)] \\ E[x(n-1)x(n)] & E[x(n-1)x(n-1)] & \dots & E[x(n-1)x(n-M+1)] \\ \vdots & \vdots & \ddots & \vdots \\ E[x(n-M+1)x(n)] & E[x(n-M+1)x(n-1)] & \dots & E[x(n-M+1)x(n-M+1)] \end{pmatrix}$$

$$= \begin{pmatrix} R_{XX}(0) & R_{XX}(-1)] & \dots & R_{XX}(1-M) \\ R_{XX}(1) & R_{XX}(0)] & \dots & R_{XX}(2-M) \\ \vdots & \vdots & \ddots & \vdots \\ R_{XX}(M-1) & R_{XX}(M-2)] & \dots & R_{XX}(0)] \end{pmatrix}$$

# The Wiener filter

- Assuming that  $\mathbf{d}(n)$  has zero mean,  $E[d^2(n)] = \sigma_d^2$ , therefore

$$\text{MSE} = \sigma_d^2 - 2\vec{w}^T \overrightarrow{R_{XD}} + \vec{w}^T \mathbf{R}_{XX} \vec{w}$$

$$\begin{aligned} \overrightarrow{R_{XD}} &= E[\vec{x}(n)d(n)] = [E[x(n)d(n)], E[x(n-1)d(n)], \dots, E[x(n-(M-1))d(n)]]^T \\ &= [R_{XD}(0), R_{XD}(1), \dots, R_{XD}(M-1)]^T \end{aligned}$$

$$\mathbf{R}_{XX} = \begin{pmatrix} R_{XX}(0) & R_{XX}(-1) & \dots & R_{XX}(-(M-1)) \\ R_{XX}(1) & R_{XX}(0) & \dots & R_{XX}(-(M-2)) \\ \vdots & \vdots & \ddots & \vdots \\ R_{XX}(M-1) & R_{XX}(M-2) & \dots & R_{XX}(0) \end{pmatrix}$$

# The Wiener filter: optimal coefficients

- Let us now obtain the derivative of the MSE to find the minimum

$$\text{MSE} = \sigma_d^2 - 2\vec{w}^T \overrightarrow{R_{XD}} + \vec{w}^T \mathbf{R}_{XX} \vec{w}$$

$$\frac{d\text{MSE}}{d\vec{w}} = -2 \frac{d(\vec{w}^T \overrightarrow{R_{XD}})}{d\vec{w}} + \frac{d(\vec{w}^T \mathbf{R}_{XX} \vec{w})}{d\vec{w}}$$

$$\begin{aligned} \frac{d(\vec{w}^T \overrightarrow{R_{XD}})}{d\vec{w}} &= \left[ \frac{\partial(\vec{w}^T \overrightarrow{R_{XD}})}{\partial w_0}, \frac{\partial(\vec{w}^T \overrightarrow{R_{XD}})}{\partial w_1}, \dots, \frac{\partial(\vec{w}^T \overrightarrow{R_{XD}})}{\partial w_{M-1}} \right]^T \\ &= \left[ \frac{\partial(\sum_{i=0}^{M-1} \vec{w}(i) \overrightarrow{R_{XD}}(i))}{\partial w_0}, \frac{\partial(\sum_{i=0}^{M-1} \vec{w}(i) \overrightarrow{R_{XD}}(i))}{\partial w_1}, \dots, \frac{\partial(\sum_{i=0}^{M-1} \vec{w}(i) \overrightarrow{R_{XD}}(i))}{\partial w_{M-1}} \right]^T \\ &= [\overrightarrow{R_{XD}}(0), \overrightarrow{R_{XD}}(1), \dots, \overrightarrow{R_{XD}}(M-1)]^T \\ &= [R_{XD}(0), R_{XD}(1), \dots, R_{XD}(M-1)]^T = \overrightarrow{R_{XD}} \end{aligned}$$

# The Wiener filter: optimal coefficients

$$\frac{dMSE}{d\vec{w}} = -2 \frac{d(\vec{w}^T \overrightarrow{R_{XD}})}{d\vec{w}} + \frac{d(\vec{w}^T \mathbf{R}_{XX} \vec{w})}{d\vec{w}} = -2\overrightarrow{R_{XD}} + \frac{d(\vec{w}^T \mathbf{R}_{XX} \vec{w})}{d\vec{w}}$$


---

$$\frac{d(\vec{w}^T \mathbf{R}_{XX} \vec{w})}{d\vec{w}} = \frac{d(\sum_{i=0}^{M-1} w_i \sum_{j=0}^{M-1} w_j R_{XX}(j-i))}{d\vec{w}}$$

$$= \left[ \begin{array}{cccc} \sum_{i=0}^{M-1} w_i R_{XX}(i) + \sum_{j=0}^{M-1} w_j R_{XX}(-j), & \sum_{i=0}^{M-1} w_i R_{XX}(i-1) + \sum_{j=0}^{M-1} w_j R_{XX}(1-j), \\ \dots, & \sum_{i=0}^{M-1} w_i R_{XX}(i-(M-1)) + \sum_{j=0}^{M-1} w_j R_{XX}((M-1)-j) \end{array} \right]^T$$

$$= [\sum_{i=0}^{M-1} w_i (R_{XX}(i) + R_{XX}(-j)), \sum_{i=0}^{M-1} w_i (R_{XX}(i-1) + R_{XX}(1-j)), \dots, \sum_{i=0}^{M-1} w_i (R_{XX}(i-(M-1)) + R_{XX}((M-1)-j))]^T$$

X is stationary

$$= [2 \sum_{i=0}^{M-1} w_i R_{XX}(-i), 2 \sum_{i=0}^{M-1} w_i R_{XX}(1-i), \dots, 2 \sum_{i=0}^{M-1} w_i R_{XX}((M-1)-i)]^T$$

$$= 2\mathbf{R}_{XX}\vec{w}$$

# The Wiener filter: optimal coefficients

- To minimize the MSE we set its derivative to zero

$$\frac{dMSE}{d\vec{w}} = -2\overrightarrow{R_{XD}} + 2\mathbf{R}_{XX} \vec{w} = 0$$

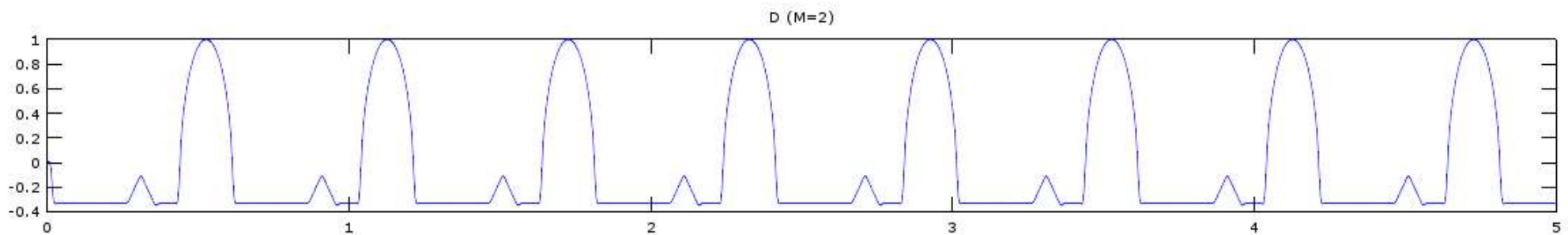
- Therefore

$$\mathbf{R}_{XX} \vec{w} = \overrightarrow{R_{XD}}$$

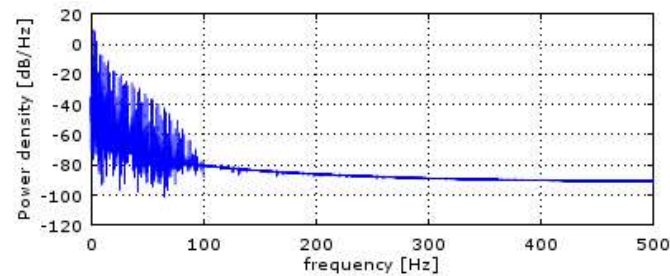
$$\vec{w} = \mathbf{R}_{XX}^{-1} \overrightarrow{R_{XD}}$$

# The Wiener filter: Case study

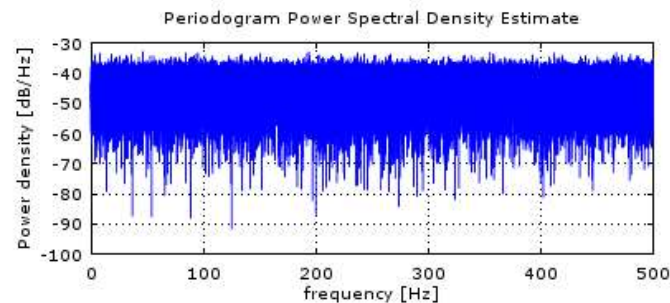
- A synthetic ECG is corrupted with white noise
- The original SNR is 10 dB
- The corrupted signal is filtered using Wiener filters with M taps (order M-1)



$S_{dd}(f)$



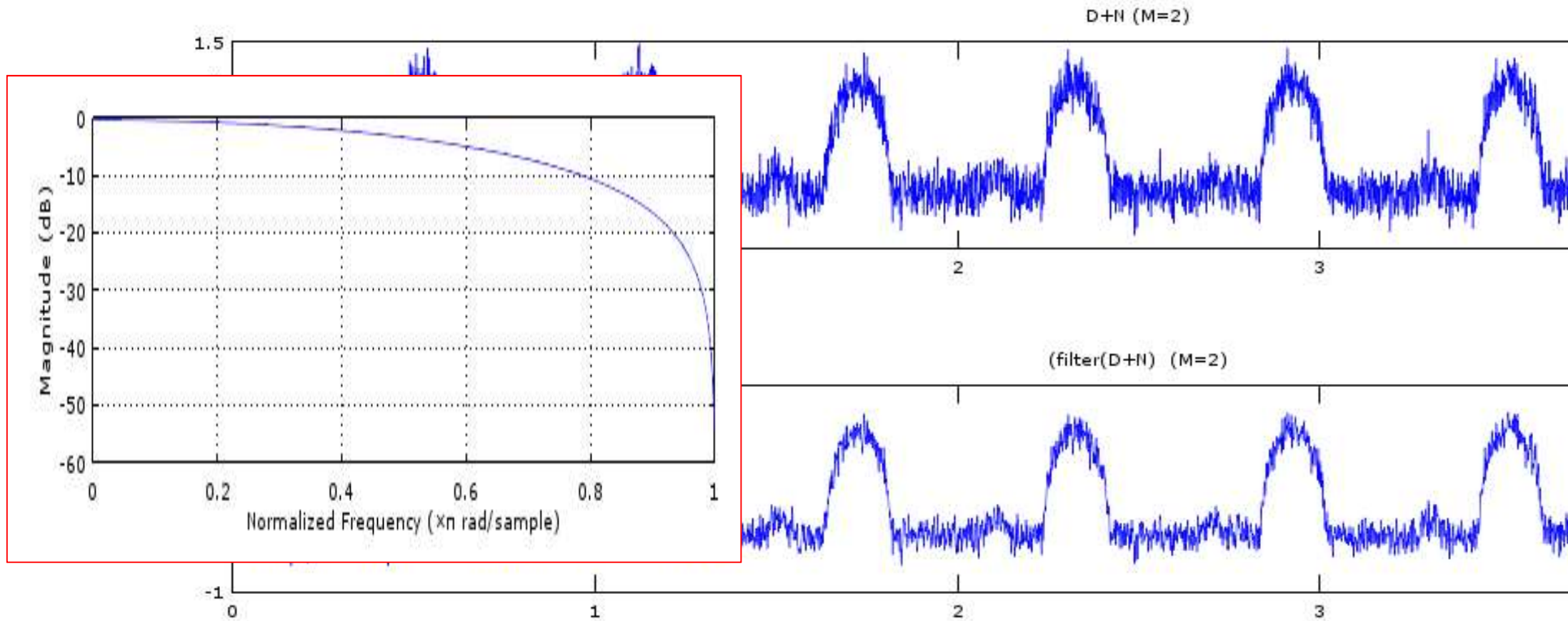
$S_{nn}(f)$



# The Wiener filter: Case study

- $M=2$  (order 1)

SNR=10 dB  $\rightarrow$  13 dB

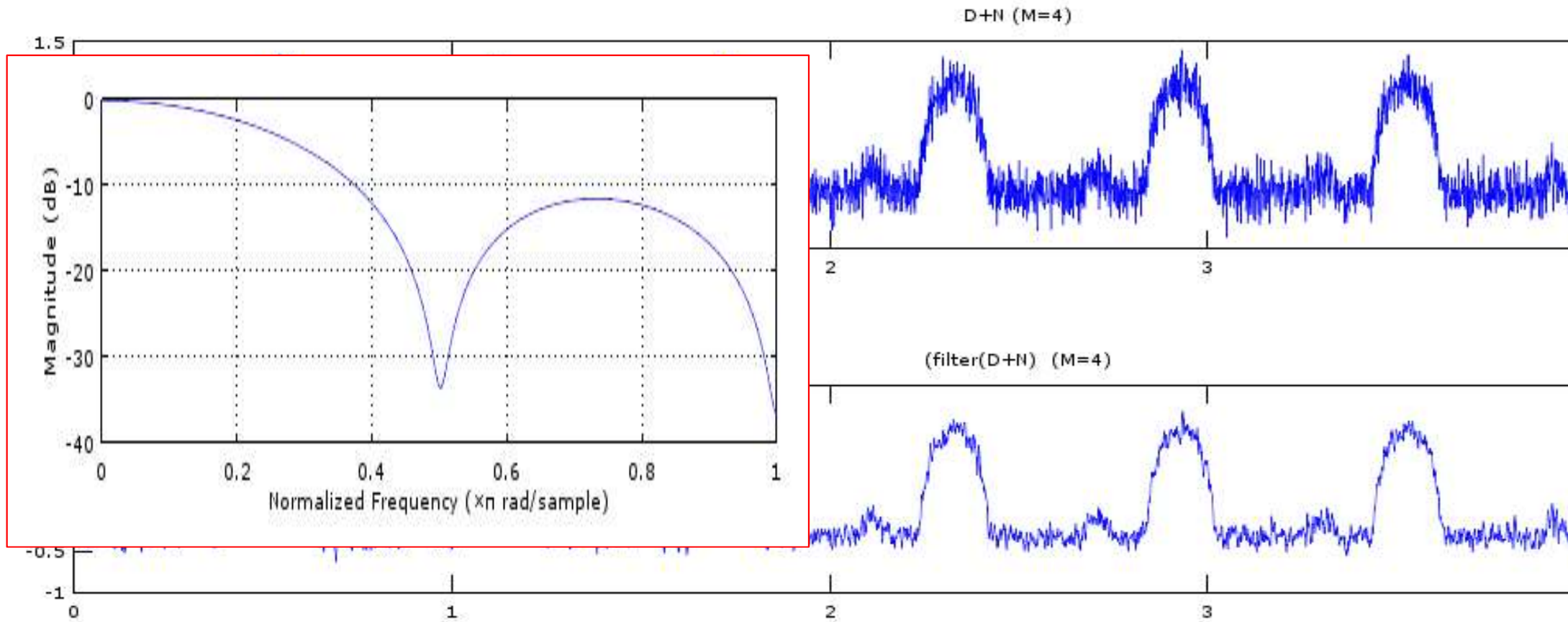




# The Wiener filter: Case study

- $M=4$  (order 3)

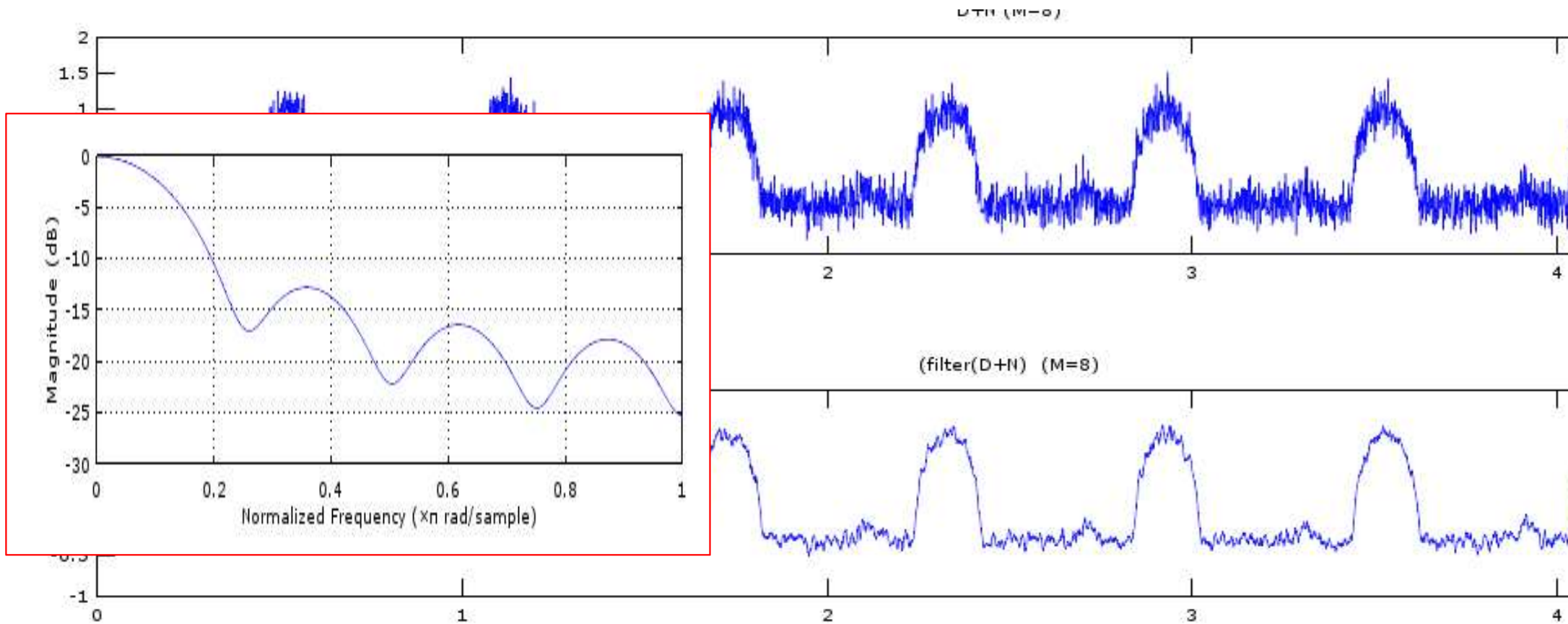
SNR=10 dB  $\rightarrow$  16 dB



# The Wiener filter: Case study

- $M=8$  (order 7)

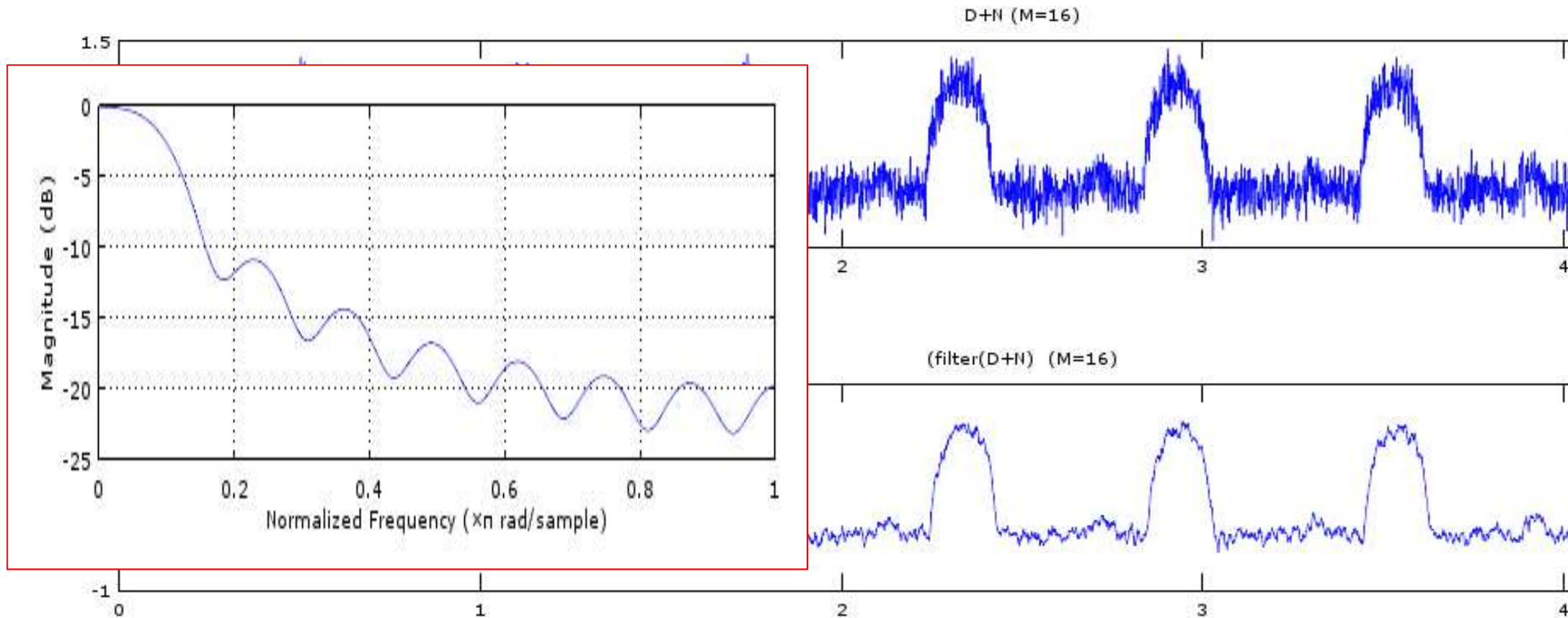
SNR=10 dB  $\rightarrow$  18.7 dB



# The Wiener filter: Case study

- M=16 (order 15)

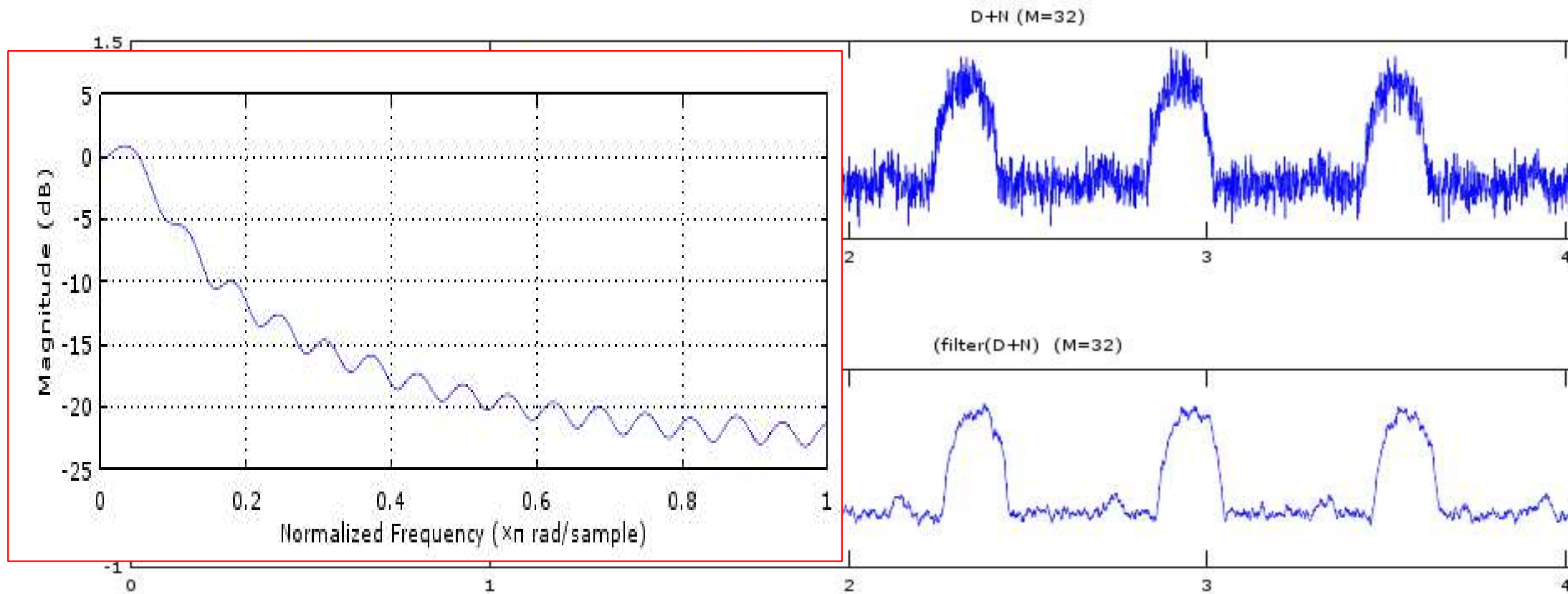
SNR=10 dB  $\rightarrow$  19.2 dB



# The Wiener filter: Case study

- $M=32$  (order 31)

SNR=10 dB  $\rightarrow$  19.5 dB



# SUMMARY

- Noise filtering
- Wiener filtering
- Wiener-Hopf equations
- ECG example